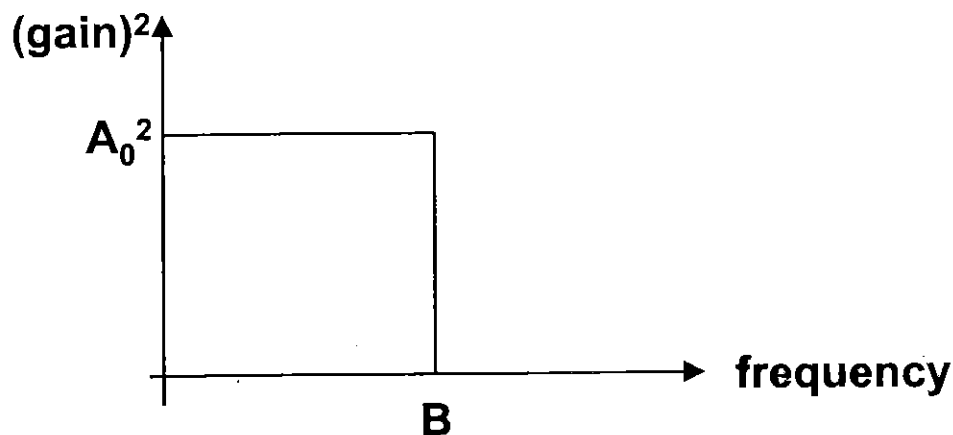
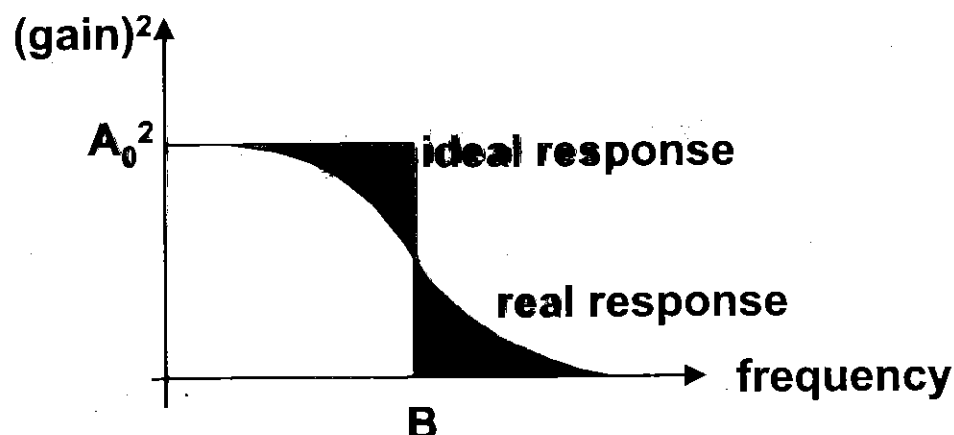


Noise Equivalent Bandwidth

- This is defined as the voltage-gain-squared bandwidth of the circuit
- The ideal case is that the $(\text{gain})^2$ is constant at a value of A_0^2 up to the bandwidth ($A_0 = \text{voltage gain}$)



- But, the behaviour of a real circuit is not abrupt



- The NEB is defined as the point at which the two shaded areas equal

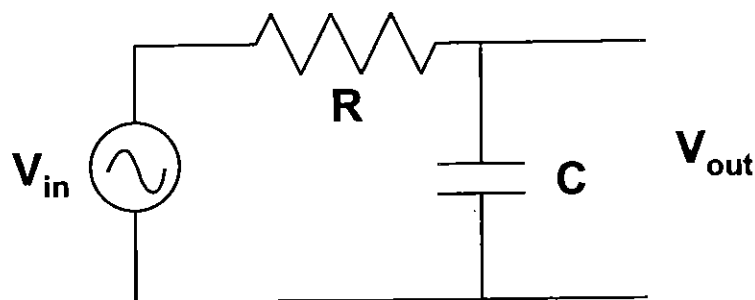
- Mathematically, this is given by

$$B = \frac{1}{|A_0|^2} \int_0^{\infty} |A(f)|^2 df$$

- So in the ideal case

$$\text{bandwidth} = \frac{1}{A_0^2} \int_0^{\infty} |A|^2 df = \frac{A_0^2 B}{A_0^2} = B$$

- If we take the example of an RC low pass filter



- Calculating the transfer function

$$\begin{aligned} A(\omega) &= \frac{V_{\text{out}}}{V_{\text{in}}} = \frac{\frac{1}{j\omega C}}{\frac{1}{j\omega C} + R} \\ &= \frac{1}{1 + 2\pi fRC} \quad \text{since } \omega = 2\pi f \\ &= \frac{f_0}{jf + f_0} \quad \text{where } f_0 = \frac{1}{2\pi RC} \end{aligned}$$

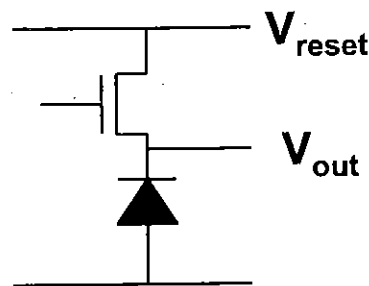
- At $f = 0$, $A(f) = A_0 = 1$ for this circuit
- Now we can calculate the noise equivalent bandwidth, using $A_0 = 1$

$$\begin{aligned}
 B &= \int_0^{\infty} \left(\frac{f_0}{\sqrt{f^2 + f_0^2}} \right)^2 df \\
 &= f_0^2 \int_0^{\infty} (f_0^2 + f^2)^{-1} df \\
 &= \frac{\pi}{2} f_0
 \end{aligned}$$

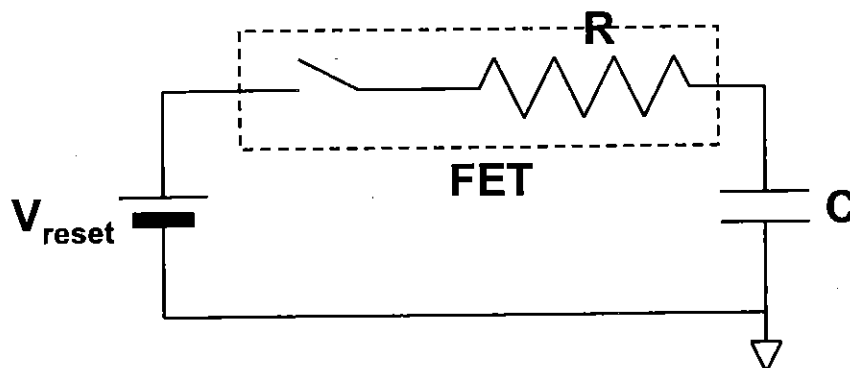
- The reason for choosing this example is that it is directly applicable to the resetting of photodiodes and the output nodes of CCD and photogate pixels

Reset Noise

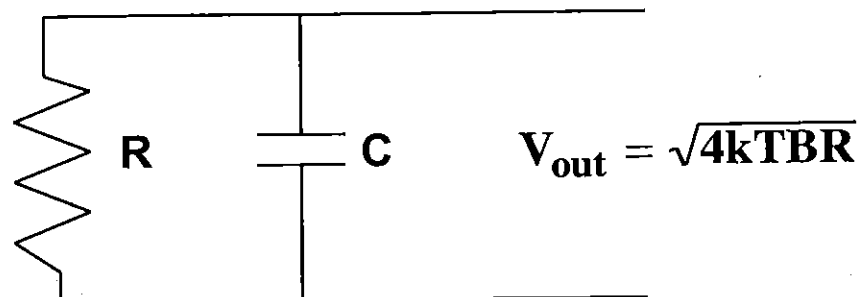
- If we consider a diffusion (either a floating diffusion or a photodiode) being reset through a MOSFET



- Effectively, this is a capacitance being charged through the resistance of the MOSFET channel



- So the ac-equivalent circuit is



- From before, the bandwidth is

$$B = \frac{\pi}{2} f_0 = \frac{1}{4RC}$$

- So we find the rms noise voltage

$$\langle V_{out} \rangle = \frac{1}{\sqrt{2}} \sqrt{\frac{kT}{C}}$$

- Usually, the noise voltages is expressed in terms of electrons, in order to compare directly with the electrons in the well
- In which case the reset noise on the capacitor is calculated from $Q = nq = Cv_{out}$, and the rms noise electrons is given by

$$\langle n_{e-} \rangle = \frac{C}{q} \sqrt{\frac{kT}{C}} = \sqrt{\frac{kTC}{q^2}}$$

- This noise is generally called "kTC noise" or, in this case, reset noise
- Calculating this out at room temperature gives

$$\langle n_{kTC,RT} \rangle = 400 \sqrt{C(pF)}$$
- For a floating diffusion $C \sim 20fF$, so $n_{kTC} = 55 e^-$
- For a $(10\mu m)^2$ photodiode, $C \sim 60pF$, so $n_{kTC} = 100 e^-$

» currently, reset noise limits the read noise in PDs